

EECS150

Section 5 Simplification and State Minimization Fall 2001



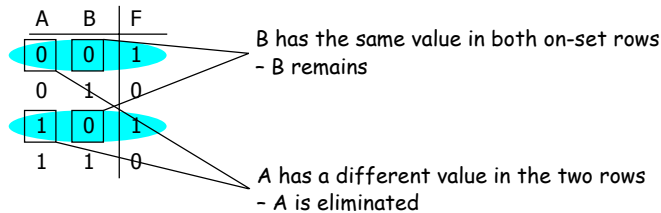
Simplification

- Finding a minimal sum of products or product of sums realization
 - Exploit don't care information in the process
- Algebraic simplification
 - Not an algorithmic/systematic procedure
 - How do you know when the minimum realization has been found?
- Computer-aided design tools
 - Precise solutions require very long computation times, especially for functions with many inputs (> 10)
 - Heuristic methods employed – "educated guesses" to reduce amount of computation and yield good if not best solutions
- Hand methods still relevant
 - To understand automatic tools and their strengths and weaknesses
 - Ability to check results (on small examples)

The Uniting Theorem

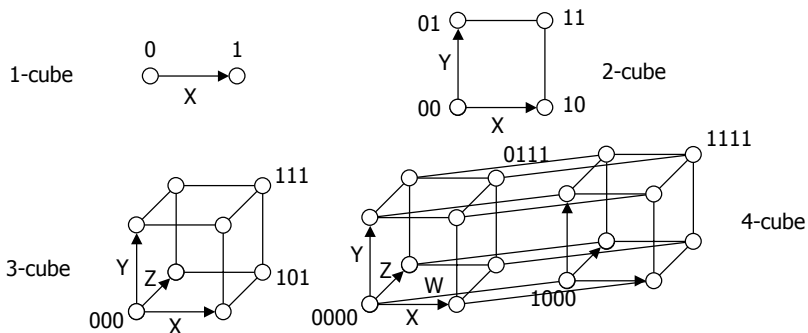
- Key tool to simplification: $A(B' + B) = A$
- Essence of simplification of two-level logic
 - Find two element subsets of the ON-set where only one variable changes its value – this single varying variable can be eliminated and a single product term used to represent both elements

$$F = A'B' + AB' = (A' + A)B' = B'$$



Boolean cubes

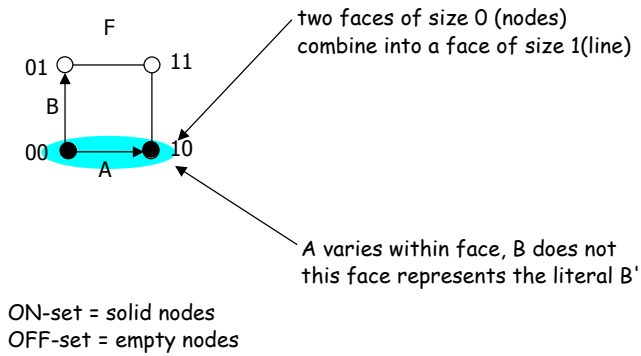
- Visual technique for indentifying when the uniting theorem can be applied
- n input variables = n-dimensional "cube"



Mapping truth tables onto cubes

- Uniting theorem combines two "faces" of a cube into a larger "face"
- Example:

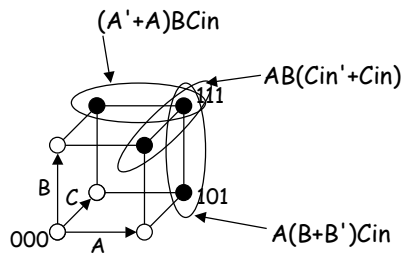
A	B	F
0	0	1
0	1	0
1	0	1
1	1	0



Three variable example

- Binary full-adder carry-out logic

A	B	Cin	Cout
0	0	0	0
0	0	1	0
0	1	0	0
0	1	1	1
1	0	0	0
1	0	1	1
1	1	0	1
1	1	1	1

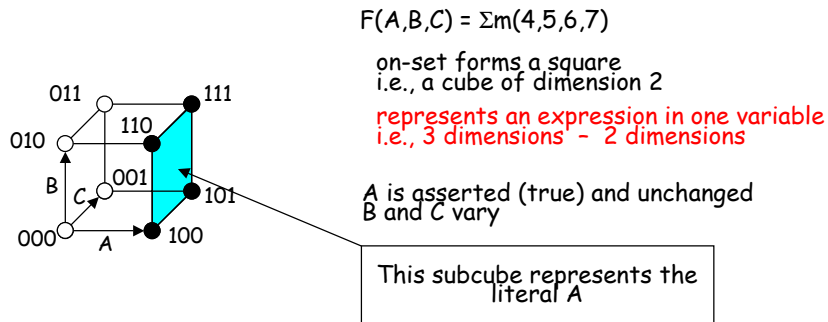


the on-set is completely covered by the combination (OR) of the subcubes of lower dimensionality - note that "111" is covered three times

$$Cout = BCin + AB + ACin$$

Higher dimensional cubes

- Sub-cubes of higher dimension than 2



m-dimensional cubes

- In a 3-cube (three variables):
 - 0-cube, i.e., a single node, yields a term in 3 literals
 - 1-cube, i.e., a line of two nodes, yields a term in 2 literals
 - 2-cube, i.e., a plane of four nodes, yields a term in 1 literal
 - 3-cube, i.e., a cube of eight nodes, yields a constant term "1"
- In general,
 - m-subcube within an n-cube ($m < n$) yields a term with $n - m$ literals

Karnaugh maps

- Flat map of Boolean cube
 - Wrap-around at edges
 - Hard to draw and visualize for more than 4 dimensions
 - Virtually impossible for more than 6 dimensions
- Alternative to truth-tables to help visualize adjacencies
 - Guide to applying the uniting theorem
 - On-set elements with only one variable changing value are adjacent unlike the situation in a linear truth-table

	A	0	1
B	0	0	1
1	0	1	2
1	1	0	3

A	B	F
0	0	1
0	1	0
1	0	1
1	1	0

Karnaugh maps (cont'd)

- Numbering scheme based on Gray-code
 - e.g., 00, 01, 11, 10
 - Only a single bit changes in code for adjacent map cells

	AB	00	01	11	10
C	0	0	2	6	4
1	1	1	3	7	5

	A	0	1
C	0	0	2
1	1	1	3

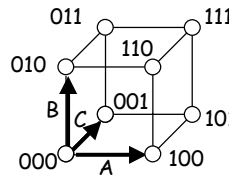
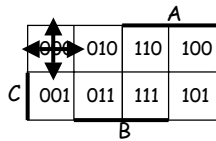
	A	0	1
C	0	0	2
1	1	1	3

	A	0	1	2	3
C	0	0	4	12	8
1	1	1	5	13	9
2	0	3	7	15	11
1	1	2	6	14	10

13 = 1101 = ABC'D

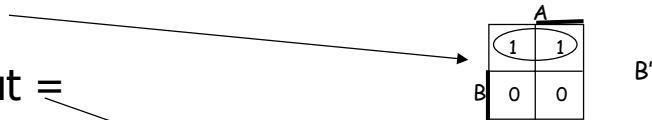
Adjacencies in Karnaugh maps

- Wrap from first to last column
- Wrap top row to bottom row

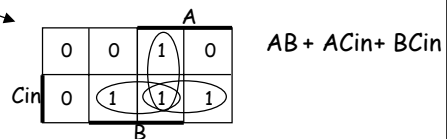


Karnaugh map examples

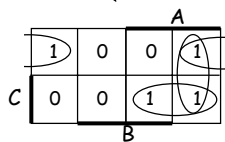
- $F =$



- $C_{out} =$



- $f(A,B,C) = \Sigma m(0,4,6,7)$



$AC + B'C' + AB'$

obtain the complement of the function by covering 0s with subcubes

More Karnaugh map examples

		A	
	0	0	1 1
C	0	0	1 1
		B	

$$G(A,B,C) = A$$

	A		
	1	0	0 1
C	0	0	1 1
		B	

$$F(A,B,C) = \sum m(0,4,5,7) = AC + B'C'$$

		A	
	0	1 1	0
C	1 1	0	0
		B	

F' simply replace 1's with 0's and vice versa

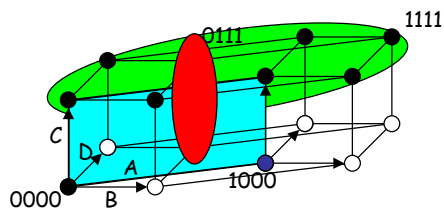
$$F'(A,B,C) = \sum m(1,2,3,6) = BC' + A'C$$

Karnaugh map: 4-variable example

- $F(A,B,C,D) = \sum m(0,2,3,5,6,7,8,10,11,14,15)$

$$F = C + A'BD + B'D'$$

		A		
	1	0	0	1
	0	1	0	0
C	1	1	1	1
	1	1	1	1
		B		
		D		



find the smallest number of the largest possible subcubes to cover the ON-set
(fewer terms with fewer inputs per term)

Karnaugh maps: don't cares

- $f(A,B,C,D) = \Sigma m(1,3,5,7,9) + d(6,12,13)$
 - without don't cares
 - $f = A'D + B'C'D$

	A			
	0	0	X	0
	1	1	X	1
C	1	1	0	0
	0	X	0	0
	B			

Karnaugh maps: don't cares

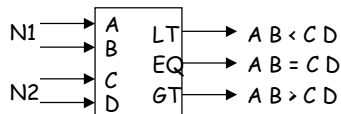
- $f(A,B,C,D) = \Sigma m(1,3,5,7,9) + d(6,12,13)$
 - $f = A'D + B'C'D$ without don't cares
 - $f = A'D + C'D$ with don't cares

	A			
	0	0	X	0
	1	1	X	1
C	1	1	0	0
	0	X	0	0
	B			

by using don't care as a "1"
a 2-cube can be formed
rather than a 1-cube to cover
this node

don't cares can be treated as
1s or 0s
depending on which is more
advantageous

Example: two-bit comparator

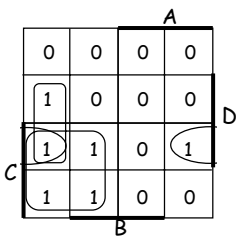


block diagram
and
truth table

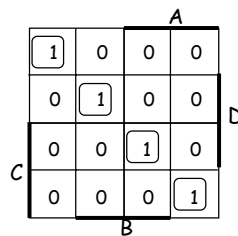
A	B	C	D	LT	EQ	GT
0	0	0	0	0	1	0
		0	1	1	0	0
		1	0	1	0	0
		1	1	1	0	0
0	1	0	0	0	0	1
		0	1	0	1	0
		1	0	1	0	0
		1	1	1	0	0
1	0	0	0	0	0	1
		0	1	0	0	1
		1	0	0	1	0
		1	1	1	0	0
1	1	0	0	0	0	1
		0	1	0	0	1
		1	0	0	0	1
		1	1	0	1	0

we'll need a 4-variable Karnaugh map
for each of the 3 output functions

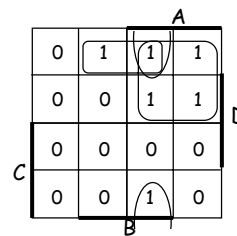
Example: two-bit comparator



K-map for LT



K-map for EQ



K-map for GT

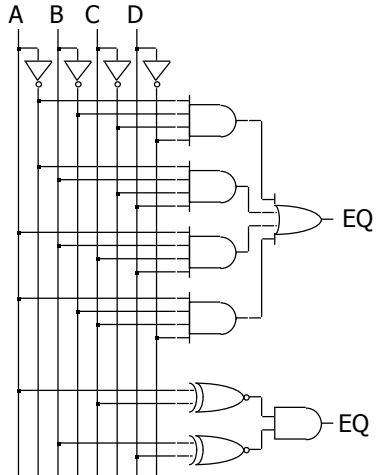
$$LT = A' B' D + A' C + B' C D$$

$$EQ = A' B' C' D' + A' B C' D + A B C D + A B' C D' = (A \text{ xnor } C) \cdot (B \text{ xnor } D)$$

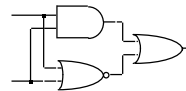
$$GT = B C' D' + A C' + A B D'$$

LT and GT are similar (flip A/C and B/D)

Example: two-bit comparator

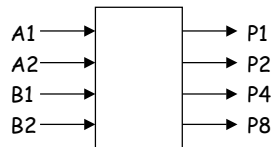


two alternative implementations of EQ with and without XOR



XNOR is implemented with at least 3 simple gates

Example: 2x2-bit multiplier

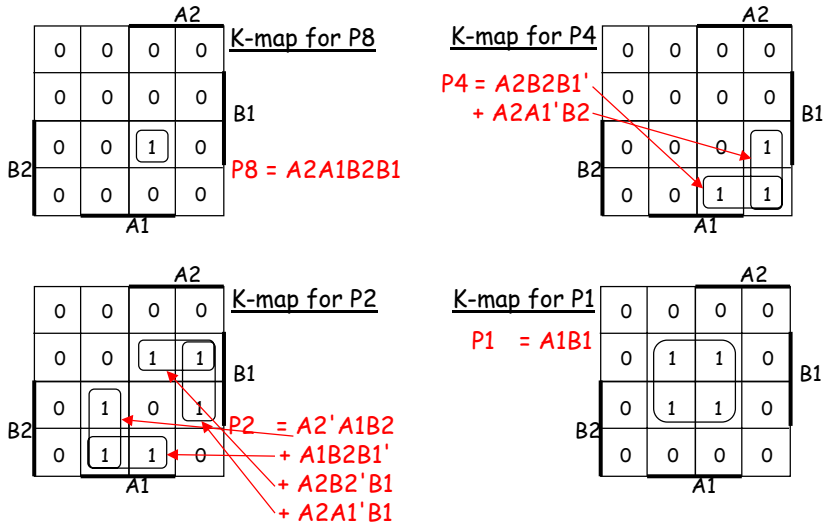


block diagram and truth table

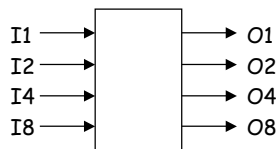
A2	A1	B2	B1	P8	P4	P2	P1
0	0	0	0	0	0	0	0
		0	1	0	0	0	0
		1	0	0	0	0	0
		1	1	0	0	0	0
0	1	0	0	0	0	0	0
		0	1	0	0	0	1
		1	0	0	0	1	0
		1	1	0	0	1	1
1	0	0	0	0	0	0	0
		0	1	0	0	1	0
		1	0	0	1	0	0
		1	1	0	1	1	0
1	1	0	0	0	0	0	0
		0	1	0	0	1	1
		1	0	0	1	1	0
		1	1	1	0	0	1

4-variable K-map for each of the 4 output functions

Example: 2x2-bit multiplier



Example: BCD increment by 1

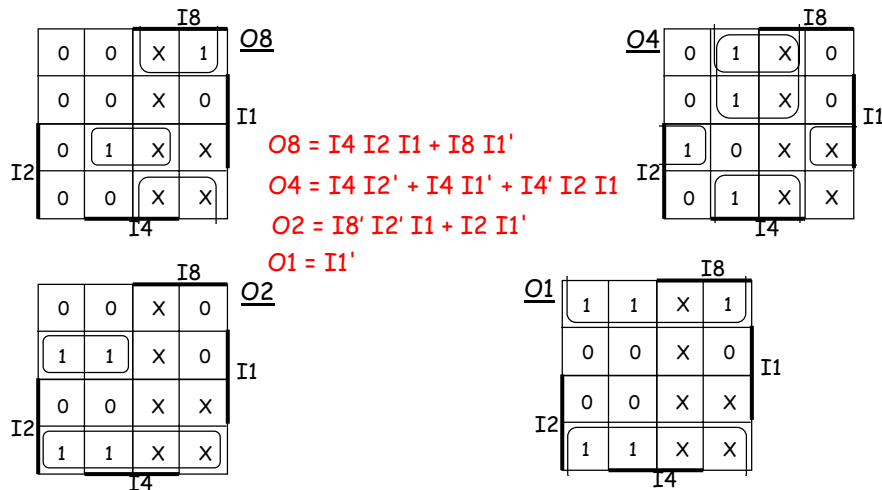


block diagram and truth table

I8	I4	I2	I1	O8	O4	O2	O1
0	0	0	0	0	0	0	1
0	0	0	1	0	0	1	0
0	0	1	0	0	0	1	1
0	0	1	1	0	1	0	0
0	1	0	0	0	0	1	0
0	1	0	1	0	1	1	1
0	1	1	1	1	0	0	0
1	0	0	0	1	0	0	0
1	0	0	1	X	X	X	X
1	0	1	1	X	X	X	X
1	1	0	0	X	X	X	X
1	1	0	1	X	X	X	X
1	1	1	0	X	X	X	X
1	1	1	1	X	X	X	X

4-variable K-map for each of the 4 output functions

Example: BCD increment by 1



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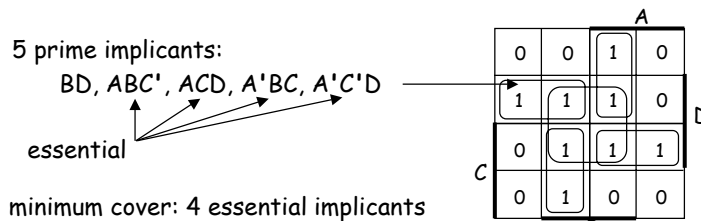
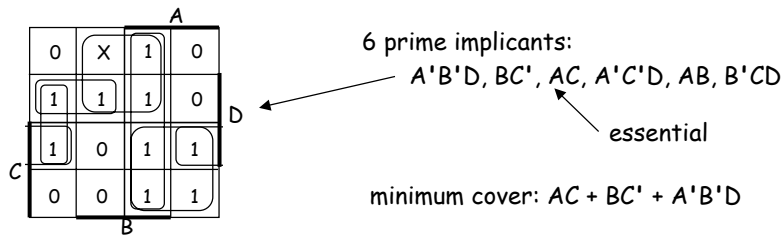
Definition of terms

- Implicant
 - Single element of ON-set or any group of these elements that can be combined to form a subcube
- Prime implicant
 - Implicant that can't be combined with another to form a larger subcube
- Essential prime implicant
 - Prime implicant is essential if it alone covers an element of ON-set
 - Will participate in ALL possible covers of the ON-set
- Objective:
 - Grow implicant into prime implicants (minimize literals per term)
 - Cover the ON-set with as few prime implicants as possible (minimize number of product terms)

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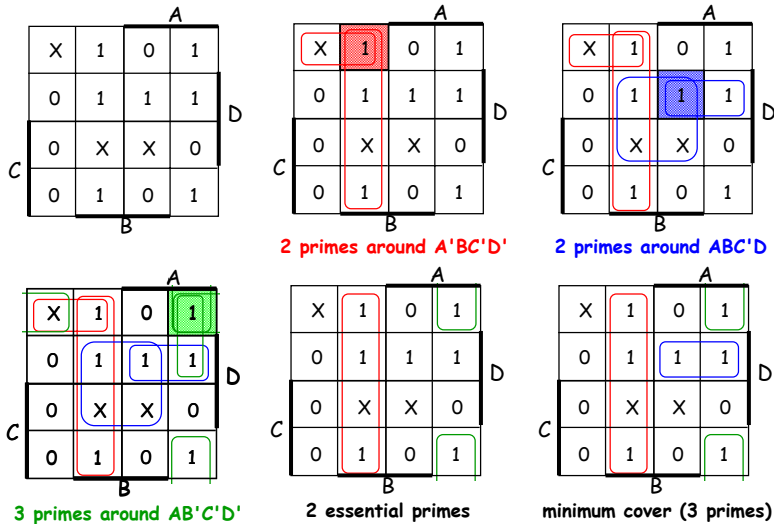
Examples to illustrate terms



Two-level simplification algorithm

- Algorithm: minimum sum-of-products expression from a Karnaugh map
 - Step 1: choose an element of the ON-set
 - Step 2: find "maximal" groupings of 1s and Xs adjacent to that element
 - consider top/bottom row, left/right column, and corner adjacencies
 - this forms prime implicants (number of elements are power of 2)
 - Repeat Steps 1 and 2 to find all prime implicants
 - Step 3: revisit the 1s in the K-map
 - if covered by single prime implicant, it is essential, and participates in final cover
 - 1s covered by essential prime implicant do not need to be revisited
 - Step 4: if there remain 1s not covered by essential prime implicants
 - select the smallest number of prime implicants that cover the remaining 1s

Example



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Finite State Machine Optimization

- State Minimization
 - Fewer states require fewer state bits
 - Fewer bits require fewer logic equations
- Encodings: State, Inputs, Outputs
 - State encoding with fewer bits has fewer equations to implement
 - However, each may be more complex
 - State encoding with more bits (e.g., one-hot) has simpler equations
 - Complexity directly related to complexity of state diagram
 - Input/output encoding may or may not be under designer control

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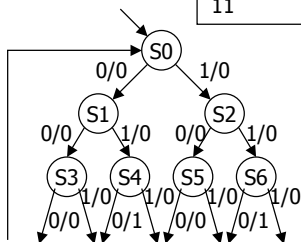
Algorithmic Approach

- Goal – identify and combine states that have equivalent behavior
- Equivalent States:
 - Same output
 - For all input combinations, states transition to same or equivalent states
- Algorithm Sketch
 1. Place all states in one set
 2. Initially partition set based on output behavior
 3. Successively partition resulting subsets based on next state transitions
 4. Repeat (3) until no further partitioning is required
 - states left in the same set are equivalent
- Polynomial time procedure

State Minimization Example

- Sequence Detector for 010 or 110

Input Sequence	Present State	Next State		Output	
		X=0	X=1	X=0	X=1
Reset	S0	S1	S2	0	0
0	S1	S3	S4	0	0
1	S2	S5	S6	0	0
00	S3	S0	S0	0	0
01	S4	S0	S0	1	0
10	S5	S0	S0	0	0
11	S6	S0	S0	1	0



Method of Successive Partitions

Input Sequence	Present State	Next State		Output	
		X=0	X=1	X=0	X=1
Reset	S0	S1	S2	0	0
0	S1	S3	S4	0	0
1	S2	S5	S6	0	0
00	S3	S0	S0	0	0
01	S4	S0	S0	1	0
10	S5	S0	S0	0	0
11	S6	S0	S0	1	0

(S0 S1 S2 S3 S4 S5 S6)

S1 is equivalent to S2

(S0 S1 S2 S3 S5) (S4 S6)

S3 is equivalent to S5

(S0 S3 S5) (S1 S2) (S4 S6)

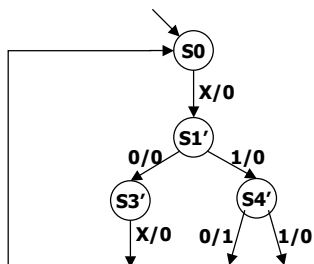
S4 is equivalent to S6

(S0) (S3 S5) (S1 S2) (S4 S6)

Minimized FSM

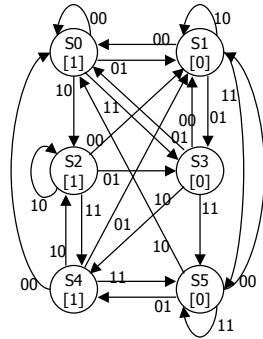
- State minimized sequence detector for 010 or 110

Input Sequence	Present State	Next State		Output	
		X=0	X=1	X=0	X=1
Reset	S0	S1'	S1'	0	0
0 + 1	S1'	S3'	S4'	0	0
X0	S3'	S0	S0	0	0
X1	S4'	S0	S0	1	0



More Complex State Minimization

Multiple input example



inputs here

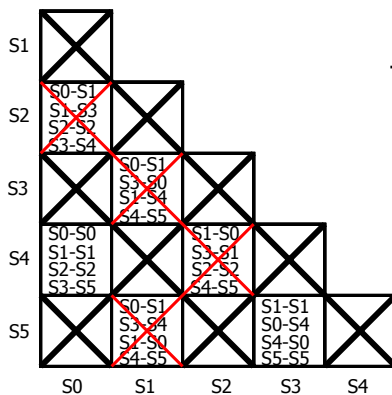
present state	next state				output
	00	01	10	11	
S0	S0	S1	S2	S3	1
S1	S0	S3	S1	S4	0
S2	S1	S3	S2	S4	1
S3	S1	S0	S4	S5	0
S4	S0	S1	S2	S5	1
S5	S1	S4	S0	S5	0

symbolic state transition table

Minimized FSM

Implication Chart Method

- Cross out incompatible states based on outputs
- Then cross out more cells if indexed chart entries are already crossed out



present state	next state				output
	00	01	10	11	
S0'	S0'	S1	S2	S3'	1
S1	S0'	S3'	S1	S3'	0
S2	S1	S3'	S2	S0'	1
S3'	S1	S0'	S0'	S3'	0

minimized state table
(S0==S4) (S3==S5)

Incompletely Specified FSMs

- Equivalence of states is transitive when machine is fully specified
- But its not transitive when don't cares are present

e.g., state output

S0	– 0	S1 is compatible with both S0 and S2 but S0 and S2 are incompatible
S1	1 –	
S2	– 1	

- No polynomial time algorithm exists for determining best grouping of states into equivalent sets that will yield the smallest number of final states